

# Low Velocity HAL SAW Resonator Using LiNbO<sub>3</sub> Thin Plate on Quartz Substrate

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**Abstract**— This paper reports a low velocity hetero acoustic layer (HAL) surface acoustic wave (SAW) resonator using a combination of 11° YX lithium niobate (LN) thin plate and YX quartz (Qz) substrate with copper (Cu) interdigital transducers (IDTs). The fabricated one-port (1-port) SAW resonator attained a low effective velocity of 2416 ms<sup>-1</sup> at resonance with an impedance ratio (Z-ratio) of 80 dB and a relative bandwidth (BW) of 11%. The resonator has temperature coefficient of frequency (TCF) of -129 ppm°C<sup>-1</sup> at resonance frequency ( $f_r$ ) and -117 ppm°C<sup>-1</sup> at antiresonance frequency ( $f_a$ ). The resonator has a roughly 30% lower velocity compared with the previously developed LT/ Qz HAL SAW resonator and a roughly 40% slower than the metallized SH-SAW velocity on 11° YX LN, indicating its potential for low-band filter applications.

**Keywords**—resonators, filters, surface acoustic waves

## I. INTRODUCTION (HEADING I)

Surface acoustic wave (SAW) devices and bulk acoustic wave (BAW) devices are the two mainstream components for radio frequency front-end (RFFE) filters in mobile communication devices. Conventionally, BAW devices are considered superior at high frequencies compared to SAW devices due to their higher Q factor. However, the introduction of SAW devices on multilayered substrates with an ultrathin lithium niobate (LN) or lithium tantalate (LT) as a top layer has sparked new interest in SAW devices since they have been found to greatly improve the characteristics of SAW devices [1], [2], [3]. This new category of SAW devices is categorized as Hetero Acoustic Layer (HAL) SAW devices in the authors' group [4].

BAW devices for the platinum frequency bands (e.g., 700 MHz ~ 1 GHz) is challenging since it requires a thick film of aluminum nitride (AlN), which is easily warped during the fabrication process. On the other hand, SAW devices require a long wavelength, which results in large sizes. The increasing number of cellular frequency bands requires more RFFE filters in mobile communication devices, therefore, there is a strong demand for miniaturized SAW devices to maintain the compactness of mobile communication devices.

The direct solution for SAW device miniaturization is to use a substrate with low phase velocity, because the frequency is determined by the quotient of phase velocity and interdigital transducer (IDT) pitch (wavelength,  $\lambda$ ). Previous studies have taken advantage of the mass loading effect in SAW devices by introducing higher density IDTs to reduce the SAW velocities [5], [6], [7]. This work combines the high-performance of ultrathin LN HAL SAW devices and the mass-loading effect to develop a low-velocity HAL SAW resonator.

## II. DESIGN AND SIMULATION

Fig. 1 shows the impedance characteristic of a conventional 42°YX LT SAW resonator with aluminum (Al) IDTs ( $\lambda = 2.2 \mu\text{m}$ ,  $0.08\lambda$  thick) used in duplexers. The resonator has a phase velocity of 3903 ms<sup>-1</sup>, a relative bandwidth of 3.3% and an impedance ratio (Z-ratio) of 58 dB.

Its high velocity and narrow relative bandwidth have made it being unsuitable for low band filtering purposes.

The impedance characteristics of 42°YX LT SAW resonator with copper (Cu,  $\lambda = 3.78 \mu\text{m}$ ,  $0.06\lambda$  thick) and gold (Au,  $\lambda = 3.78 \mu\text{m}$ ,  $0.03\lambda$  thick) IDTs is shown in Fig. 2. Both resonators have a slight increase in relative bandwidth to 4.2% with an Z-ratio of 52 dB. The Cu/ LT SAW resonator and Au/ LT SAW resonator has phase velocities of 3530 ms<sup>-1</sup> and 3677 ms<sup>-1</sup>, respectively, which are still insufficient for low band filters.

The combination of ultrathin LT and quartz (Qz) substrate has realized high Z-ratio of 80 dB, low temperature of coefficient frequency (TCF) and spurious-free responses [1]. However, the yielded phase velocities are in the range of resonators in Fig. 2.

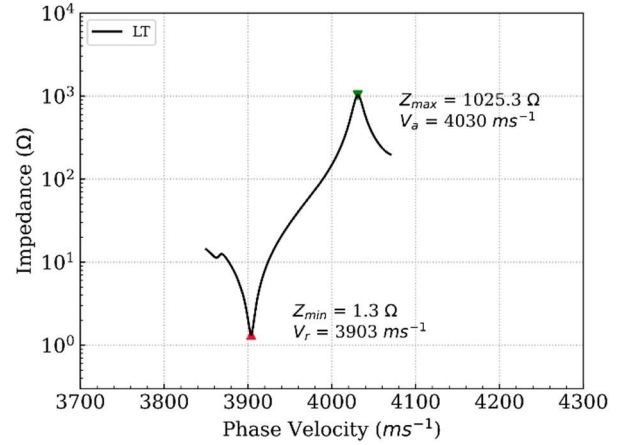


Fig. 1. Impedance characteristic of a conventional standard  $0.08\lambda$  Al/ 42°YX LT SAW resonator.

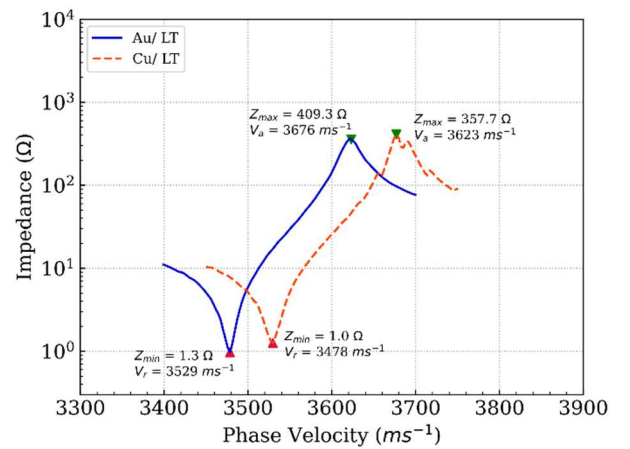


Fig. 2. Impedance characteristics of  $0.06\lambda$  Cu/ 42°YX LT SAW resonator and  $0.03\lambda$  Au/ 42°YX LT SAW resonator.

Referring to the simulation and experimental results provided by Kadota *et al.* in [1], using a LT/ SiO<sub>2</sub>/ Si multilayered structure might not be suitable due to the presence of large spurious higher-order modes. The spurious responses are caused by Sezawa wave, which is a higher-order mode of Rayleigh wave and excited by the higher velocity of Si in ZnO/ Si structure [8]. The spurious response in the LT/ SiO<sub>2</sub>/ Si multilayered structure might be analogous to the mismatch in wave velocities of the substrates.

Based on the previously mentioned results, the combination of an ultrathin LN plate and a Qz substrate was experimented in this work to obtain a suitable BW and high Z-ratio. Cu is used for the IDTs for their excellent electrical conductivity, high compatibility with LN SAW resonators reported in [9], [10], [11], and high reflection coefficient [9], which will aid in miniaturization by reducing the number of IDTs and reflector fingers [12], [13].

#### A. SH-Leaky SAW velocity

The propagation velocity of SAWs in a resonator is influenced by the material density [7] and normalized thickness [14] of the IDTs. For determining the required Cu IDT thickness, the effective SH-leaky SAW (SH-LSAW) velocities were simulated as a function of normalized Cu IDT thickness (metallization ratio, MR = 0.5) on (0°, 102.5°, 0°) LN/ (0°, 90°, 0°) Qz substrate. The results in Fig. 3 shows that a normalized Cu IDT thickness of at least 0.12λ is required to reduce effective SH-LSAW velocities below 3000 ms<sup>-1</sup>.

#### B. Impedance Ratio and Relative Bandwidth

The optimum Euler angle of the LN and Qz substrate is determined to be the angle where the 1-port SAW resonator exhibits the largest Z-ratio [1]. The Z-ratio and relative bandwidth of the 1-port LN (0.8λ thick)/ (0°, 90°, 0°) Qz HAL SAW resonator with a normalized Cu IDT thickness of 0.12λ and a MR of 0.5 were simulated as a function of LN Euler angles (0°, θ<sub>LN</sub>, 0°) in Fig. 4.

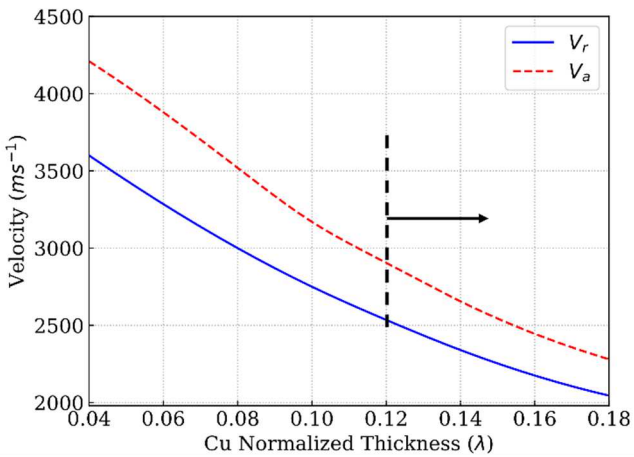


Fig. 3. Effective SH-SAW velocities as a function of normalized Cu IDT thickness (MR = 0.5) on (0°, 102.5°, 0°) LN (0.8λ thick)/ (0°, 90°, 0°) Qz substrate.  $V_r$  and  $V_a$  are phase velocities in short and open conditions, respectively.

On the other hand, the Z-ratio and relative bandwidth of the 1-port (0°, 102.5°, 0°) LN (0.8λ thick)/ Qz HAL SAW resonator with the same Cu IDT thickness and MR were simulated as a function of Qz Euler angles (0°, θ<sub>Qz</sub>, 0°) in Fig. 5. Here, the mechanical loss 1/ $Q_m$  of LN was assumed to be 0.005.

From the simulated Z-ratio, the optimum Euler angle for the LN substrate is from 100° to 103°, while the optimum Qz substrate is found to be around 90°.

### III. FABRICATION

1-port SAW resonators were fabricated on an 11° YX LN substrate and a 12.5° YX LN (~1.7 μm)/ YX Qz HAL substrate. The HAL substrate was fabricated using wafer direct bonding process, while IDTs and wiring of the device were fabricated using lift-off process [1].

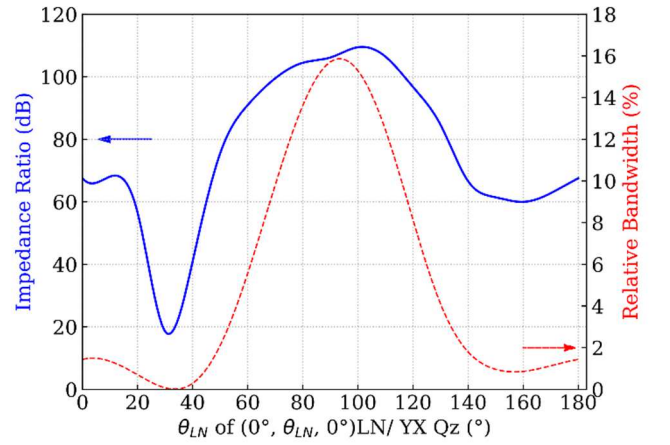


Fig. 4. Z-ratio and relative bandwidth against of θ<sub>LN</sub> of Cu (0.12λ thick, MR = 0.5)/ (0°, θ<sub>LN</sub>, 0°) LN (0.8λ thick)/ (0°, 90°, 0°) Qz resonator.

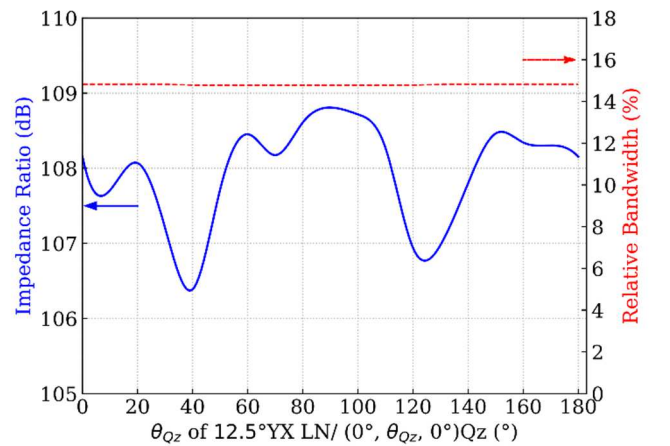


Fig. 5. Z-ratio and relative bandwidth against θ<sub>Qz</sub> of Cu (0.12λ thick, MR = 0.5)/ (0°, 102.5°, 0°) LN (0.8λ thick)/ (0°, θ<sub>Qz</sub>, 0°) Qz resonator.

#### IV. EVALUATION

The 1-port S-parameters were evaluated using Keysight's E5071C ENA Vector Network Analyzer equipped with Cascade Microtech's ACP40-GS-250 RF probe. The impedance characteristics of the  $0.065\lambda$  Cu/  $11^\circ$  YX LN and  $0.12\lambda$  Cu/  $12.5^\circ$  YX LN/ YX Qz SAW resonator are shown in Fig. 6. The LN SAW resonator has an effective velocity of  $3078 \text{ ms}^{-1}$  at resonance and achieved an Z-ratio of 67 dB with a relative bandwidth of 15%. The LN/ Qz HAL SAW resonator on the other hand has an effective velocity of  $2416 \text{ ms}^{-1}$  at resonance and achieved an Z-ratio of 80 dB with a relative bandwidth of 11%.

The frequency drift due to temperature change of the resonators are shown in Fig. 7. The  $11^\circ$  YX LN SAW resonator has a TCF of  $-129 \text{ ppm}^\circ\text{C}^{-1}$  at resonant frequency and  $-135 \text{ ppm}^\circ\text{C}^{-1}$  at anti-resonant frequency, while the  $12.5^\circ$  YX LN/ YX Qz HAL SAW resonator has a TCF of  $-129 \text{ ppm}^\circ\text{C}^{-1}$  at resonant frequency and  $-117 \text{ ppm}^\circ\text{C}^{-1}$  at anti-resonant frequency.

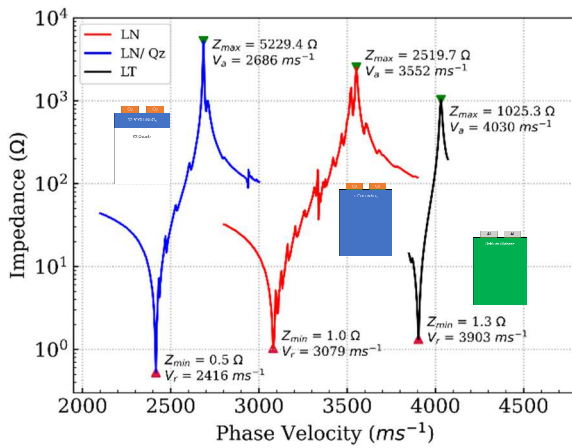


Fig. 6. Impedance characteristics of  $11^\circ$  YX LN SAW resonator and  $12.5^\circ$  YX LN/ YX Qz HAL SAW resonator.

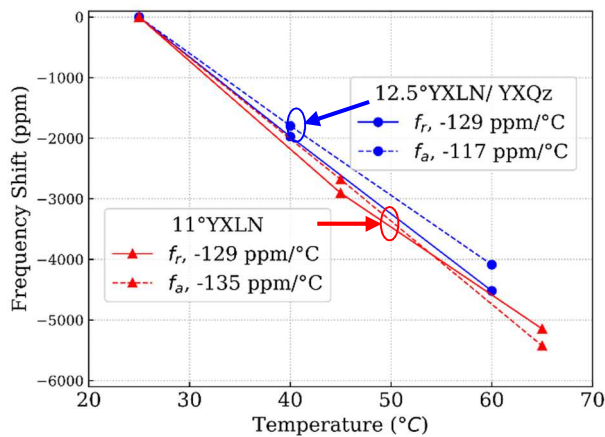


Fig. 7. Temperature drift of  $11^\circ$  YX LN SAW resonator and  $12.5^\circ$  YX LN/ YX Qz HAL SAW resonator.

#### V. CONCLUSION

The ultrathin LN and Qz HAL SAW resonator attained a low effective velocity of  $2416 \text{ ms}^{-1}$  at resonance and an Z-ratio of 80 dB. Compared to the conventional Al/  $42^\circ$  YX LT SAW resonator, the phase velocity was 40% smaller and the Z-ratio increased by 22 dB. On the other hand, using a Qz substrate with minus TCF Euler angle has resulted in a large TCF value of  $-129 \text{ ppm}^\circ\text{C}^{-1}$  and  $-117 \text{ ppm}^\circ\text{C}^{-1}$ . It is expected that the TCF values can be further lowered by substituting the YX Qz with a Qz with positive TCF Euler angle or with materials with lower coefficient of linear thermal expansion (CTE).

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